RICOH

R1232D SERIES

PWM STEP-DOWN DC/DC CONVERTER WITH SYNCHRONOUS RECTIFIER

NO.EA-129-130510

OUTLINE

The R1232D Series are CMOS-based PWM step-down DC/DC converters with synchronous rectifier, low supply current. As an output capacitor, a 10µF or more ceramic capacitor can be used with the R1232D.

Each of these ICs consists of an oscillator, a PWM control circuit, a voltage reference unit, an error amplifier, a soft-start circuit, protection circuits, a protection against miss operation under low voltage (UVLO), a chip enable circuit, a synchronous rectifier, Nch. driver transistor, and so on. A low ripple, high efficiency step-down DC/DC converter can be easily composed of this IC with only a few kinds of external components, or an inductor and capacitors. (As for R1232D001x type, divider resistors are also necessary.) In terms of the output voltage, it is fixed internally in the R1232Dxx1x types. While in the R1232D001x types, the output voltage is adjustable with external divider resistors.

As protection circuits, current limit circuit which limits peak current of Lx at each clock cycle, and latch type protection circuit exist. The latch protection works if the term of the over-current condition keeps on a certain time. Latch-type protection circuit works to latch an internal driver with keeping it disable. To release the condition of protection, after disable this IC with a chip enable circuit, enable it again, or restart this IC with power-on or make the supply voltage at UVLO detector threshold level or lower than UVLO.

FEATURES

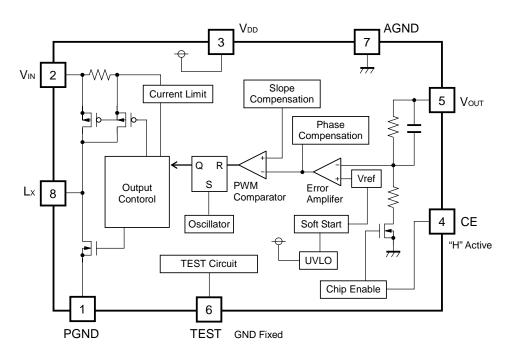
 Two choices of Oscillator Frequen 	cy1MHz, 2.25MHz
	(Small inductors can be used. 4.7μH for 1MHz/2.2μH for 2.25MHz)
Built-in Driver ON Resistance	P-channel 0.2Ω (at V _{IN} =5.0V)
Built-in Soft-start Function	Typ. 1.0ms (fosc=1MHz type)
Output Voltage	0.9V to 3.3V (R1232Dxx1x Type)
	0.8V to V _{IN} (R1232D001x Type)
High Accuracy Output Voltage	±2.0%
Built-in Current Limit Circuit	Typ. 1.4A
Package	SON-8 (t=0.9mm)

APPLICATIONS

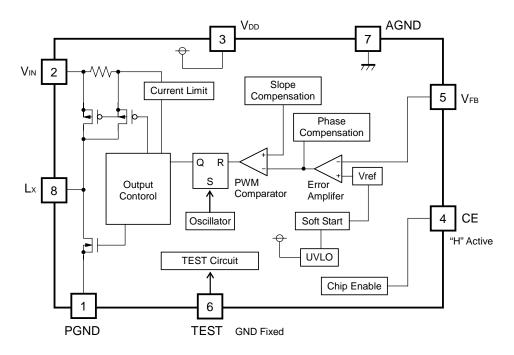
- Power source for portable equipment such as PDA, DSC, Notebook PC.
- Power source for HDD

BLOCK DIAGRAMS

R1232Dxx1A/B



R1232D001C/D



SELECTION GUIDE

In the R1232D Series, the output voltage, the oscillator frequency and the output voltage adjustment for the ICs can be selected at the user's request.

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
R1232Dxx1*-TR-FE	SON-8	3,000 pcs	Yes	Yes

xx: The output voltage can be designated in the range from 0.9 V(09) to 3.3V(33) in 0.1V steps. (For externally adjustable output voltage type, (00).)

* : The oscillator frequency and the output voltage adjustment are options as follows.

Code	Oscillator frequency	Output voltage adjustment
Α	1MHz	No
В	2.25MHz	No
С	1MHz	Yes
D	2.25MHz	Yes

PIN CONFIGURATION

Top View 8 7 6 5 1 2 3 4 Bottom View 5 6 7 8 4 3 2 1

PIN DESCRIPTIONS

Pin No	Symbol	Pin Description
1	PGND	Ground Pin
2	Vin	Voltage Supply Pin
3	V _{DD}	Voltage Supply Pin
4	CE	Chip Enable Pin (active with "H")
5	Vout/V _{FB}	Output/Feedback Pin
6	TEST	Test Pin (Forced to the GND level.)
7	AGND	Ground Pin
8	Lx	Lx Switching Pin (CMOS Output)

^{*} Tab is GND level. (They are connected to the reverse side of this IC.) The tab is better to be connected to the GND, but leaving it open is also acceptable.

ABSOLUTE MAXIMUM RATINGS

(AGND=PGND=0V)

Symbol	Item	Rating	Unit
Vin	V _{IN} Supply Voltage	-0.3 to 6.5	V
V_{DD}	V _{DD} Pin Voltage	-0.3 to 6.5	V
V_{LX}	Lx Pin Voltage	-0.3 to V _{IN} + 0.3	V
Vce	CE Pin Input Voltage	-0.3 to V _{IN} + 0.3	V
VTEST	TEST Pin Input Voltage	-0.3 to V _{IN} + 0.3	V
Vout/Vfb	Vout/VfB Pin Input Voltage	-0.3 to V _{IN} + 0.3	V
ILX	Lx Pin Output Current	±1.5	V
Po	Power Dissipation (SON-8)*	480	mW
Та	Operating Temperature Range	-40 to 85	°C
Tstg	Storage Temperature Range	-55 to 125	°C

^{*)} For Power Dissipation, please refer to PACKAGE INFORMATION.

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

ELECTRICAL CHARACTERISTICS

R1232DxxxA/C

Ta=25°C

Symbol	Item	Conditions	Min.	Тур.	Max.	Unit
Vin	Operating Input Voltage		2.6		5.5	V
Vout	Step-down Output Voltage	VIN=VCE=5.0V, IOUT=10mA	×0.980		×1.020	V
V _{FB}	Feedback Voltage	R1232D001C	0.784	0.800	0.816	V
ΔVουτ/ΔΤα	Step-down Output Voltage Temperature Coefficient	-40°C ≦ Ta ≦ 85°C		±150		ppm/ °C
fosc	Oscillator Frequency	VIN=VCE =VSET +1.5V	0.75	1.00	1.25	MHz
IDD	Supply Current	VIN=VCE =5.5V, VOUT(VFB)=5.5V	70	140	190	μΑ
Istandby	Standby Current	VCE=VOUT(VFB)=0V, VIN= 5.5V		0.0	5.0	μΑ
LXleak	Lx Leakage Current	VIN=5.5V,VCE=0V VLX=5.5V or 0V	-5.0	0.0	5.0	μА
RONP	ON Resistance of Pch Transistor	VIN=5.0V, ILX=200mA		0.20	0.35	Ω
Ronn	ON Resistance of Nch Transistor	VIN=5.0V, ILX=200mA		0.20	0.35	Ω
Maxduty	Oscillator Maximum Duty Cycle		100			%
tstart	Soft-start Time	VIN=VCE =5.0V, at no load	0.5	1.0	1.4	ms
tprot	Protection Delay Time	VIN=VCE =5.0V	0.1	2.0	10.0	ms
l∟xlimit	Lx Current Limit	VIN=VCE =5.0V	1.0	1.4		Α
Vuvlo1	UVLO Detector Threshold	VIN=VCE =2.6V-> 1.5V	2.10	2.25	2.40	V
Vuvlo2	UVLO Released Voltage	VIN=VCE =1.5V-> 2.6V	2.20	V _{UVLO1} +0.10	2.50	V
ICE	CE Input Current	VIN=5.5V, VCE =5.5V or 0V	-0.1	0.0	0.1	μΑ
Ivouт (Ivfв)	Vouт/Ivгв Leakage Current	VIN=5.5V, VCE =0V, VOUT(IVFB)=5.5V or 0V	-0.1	0.0	0.1	μА
Vceh	CE "H" Input Voltage	Vin=5.5V	1.5			V
VCEL	CE "L" Input Voltage	V _{IN} =3.0V			0.3	V

RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

• R1232DxxxB/D

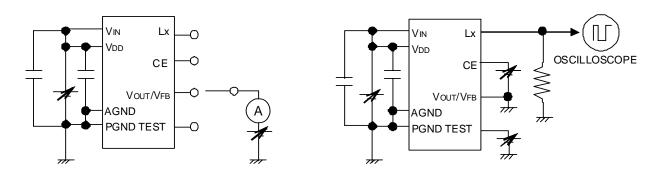
Ta=25°C

Symbol	Item	Conditions	Min.	Тур.	Max.	Unit
Vin	Operating Input Voltage		2.6		5.5	V
Vouт	Step-down Output Voltage	VIN=VCE=5.0V,IOUT=10mA	×0.980		×1.020	V
V _{FB}	Feedback Voltage	R1232D001D	0.784	0.800	0.816	V
ΔVουτ/ΔΤα	Step-down Output Voltage Temperature Coefficient	-40°C ≦ Ta ≦ 85°C		±150		ppm/ °C
fosc	Oscillator Frequency	VIN=VCE=VSET+1.5V	1.91	2.25	2.58	MHz
ldd	Supply Current	VIN=VCE =5.5V, VOUT(VFB)=5.5V	170	240	310	μА
Istandby	Standby Current	VCE=VOUT(VFB)=0V, VIN= 5.5V		0.0	5.0	μΑ
LXleak	Lx Leakage Current	V _{IN} =5.5V, V _{CE} =0V, V _{LX} =5.5V or 0V	-5.0	0.0	5.0	μΑ
RONP	ON Resistance of Pch Transistor	VIN=5.0V, ILX=200mA		0.20	0.35	Ω
Ronn	ON Resistance of Nch Transistor	VIN=5.0V, ILX=200mA		0.20	0.35	Ω
Maxduty	Oscillator Maximum Duty Cycle		100			%
tstart	Soft-start Time	VIN=VCE=5.0V, at no load	0.15	0.4	0.7	ms
tprot	Protection Delay Time	VIN=VCE=5.0V	0.1	2.0	10.0	ms
l∟xlimit	Lx Current Limit	VIN=VCE=5.0V	1.0	1.4		Α
V _{UVLO1}	UVLO Detector Threshold	VIN=VCE=2.6V -> 1.5V	2.10	2.25	2.40	V
Vuvlo2	UVLO Released Voltage	VIN=VCE =1.5V -> 2.6V	2.20	V _{UVLO1} +0.10	2.50	V
I CE	CE Input Current	VIN=5.5V, VCE =5.5V/0V	-0.1	0.0	0.1	μΑ
Ivouт (Ivfв)	Vouт/Ivfв Leakage Current	VIN=5.5V, VCE = 0V, VOUT(IVFB)=5.5V or 0V	-0.1	0.0	0.1	μА
Vсен	CE "H" Input Voltage	V _{IN} =5.5V	1.5			V
Vcel	CE "L" Input Voltage	VIN=3.0V			0.3	V

RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

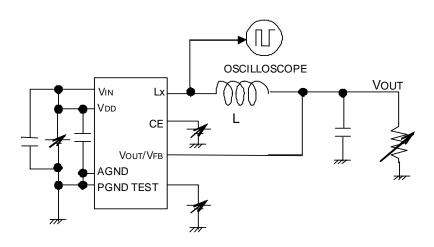
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TEST CIRCUIT

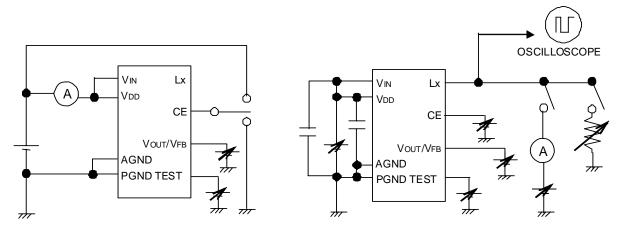


Test Circuit for Input Current and Leakage Current

Test Circuit for Input Voltage and UVLO voltage



Test Circuit for Output Voltage, Oscillator Frequency, Soft-Starting Time

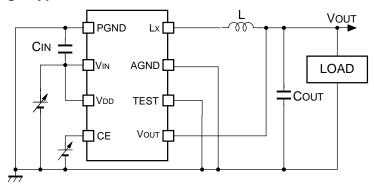


Test Circuit for Supply Current and Standby Current

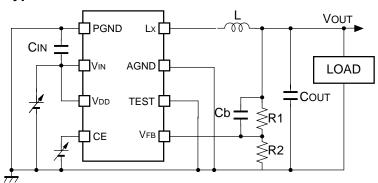
Test Circuit for ON resistance of Lx, Limit Current, Delay Time of Protection Circuit

TYPICAL APPLICATION AND TECHNICAL NOTES

• Fixed Output Voltage Type



• Adjustable Output Type



Cin	10μF C2012JB0J106MT (TDK), 10μF CM21B106M06AB (Kyocera)
Соит	10μF C2012JB0J106MT (TDK), 10μF CM21B106M06AB (Kyocera)
L	4.7μH/2.7μH VLP5610-4R7MR90, VLP5610-2R7M1R0 (TDK) *2.2μH is also suitable for B/D version.

In terms of setting R1, R2, Cb, refer to the technical notes.

When you use these ICs, consider the following issues;

- Input the same voltage into power supply pins, V_{IN} and V_{DD}. Set the same level as AGND and PGND.
- When you control the CE pin by another power supply, do not make its "H" level more than the voltage level of V_{IN} / V_{DD} pin.
- Set external components such as an inductor, C_{IN} , C_{OUT} as close as possible to the IC, in particular, minimize the wiring to V_{IN} pin and PGND pin.
- At stand by mode, (CE="L"), the LX output is Hi-Z, or both P-channel transistor and N-channel transistor of Lx pin turn off.
- Set the "Test pin" to the GND. Do not make the test pin voltage as floating or other voltage.
- Reinforce the V_{IN}, PGND, and V_{OUT} lines sufficiently. Large switching current may flow in these lines. If the impedance of V_{IN} and PGND lines is too large, the internal voltage level in this IC may shift caused by the switching current, and the operation might be unstable.
- Over current protection circuit supervises the inductor peak current (the current flowing Pch transistor) at all
 each switching cycle, and if the current beyond the Lx current limit, Pch transistor is turned off. Further, if the
 over current status continues equal or longer than protection delay time, or when the Lx limit current is
 exceeded even once when the driver operates by duty 100%, Pch transistor is latched in the OFF state and
 the operation of DC/DC converter stops.

The performance of power source circuits using these ICs extremely depends upon the peripheral circuits. Pay attention in the selection of the peripheral circuits. In particular, design the peripheral circuits in a way that the values such as voltage, current, and power of each component, PCB patterns and the IC do not exceed their respected rated values.

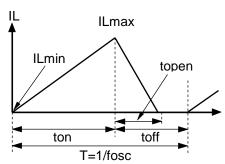
OPERATION of step-down DC/DC converter and Output Current

The step-down DC/DC converter charges energy in the inductor when Lx transistor is ON, and discharges the energy from the inductor when Lx transistor is OFF and controls with less energy loss, so that a lower output voltage than the input voltage is obtained. The operation will be explained with reference to the following diagrams:

<Basic Circuits>

In the second s

<Current through L>



Step 1:P-channel Tr. turns on and current IL (=i1) flows, and energy is charged into C_L. At this moment, IL increases from Ilmin (=0) to reach ILmax in proportion to the on-time period (ton) of P-channel Tr.

Step 2: When P-channel Tr. turns off, Synchronous rectifier N-channel Tr. turns on in order that L maintains IL at ILmax, and current IL (=i2) flows.

Step 3:IL (=i2) decreases gradually and reaches IL=ILmin=0 after a time period of topen, and N-channel Tr. Turns off. Provided that in the continuous mode, next cycle starts before IL becomes to 0 because toff time is not enough. In this case, IL value increases from this Ilmin (>0).

In the case of PWM control system, the output voltage is maintained by controlling the on-time period (ton), with the oscillator frequency (fosc) being maintained constant.

Continuous Conduction Mode

The maximum value (ILmax) and the minimum value (ILmin) of the current flowing through the inductor are the same as those when P-channel Tr. turns on and off.

The difference between ILmax and ILmin, which is represented by ΔI ;

Where, t=1/fosc=ton+toff duty (%)=ton/t×100=ton×fosc×100 topen \leq toff

In Equation 1, $V_{OUT} \times topen/L$ and $(V_{IN} - V_{OUT}) \times ton/L$ are respectively shown the change of the current at ON, and the change of the current at OFF.

Even if the output current (Iout) is, topen < toff as illustrated in the above diagram is not realized with this IC. At least, topen is equal toff (topen=toff), and when Iout is further increased, ILmin becomes larger than zero (ILmin>0). The mode is referred to as the continuous mode.

In the continuous mode, when Equation 1 is solved for ton	and assumed that the solution is tonc
tonc=t×Vout/Vin	Equation 2
When the ton=tonc, the mode is the continuous mode.	·

OUTPUT CURRENT AND SELECTION OF EXTERNAL COMPONENTS

When P-channel Tr. of Lx is ON:

(Wherein, Ripple Current P-P value is described as I_{RP} , ON resistance of P-channel Tr. and N-channel Tr. of L_X are respectively described as R_{ONP} and R_{ONN} , and the DC resistor of the inductor is described as R_L .)

VIN=VOUT+(RONP+RL)×IOUT+L×IRP/ton
When P-channel Tr. of Lx is "OFF"(N-channel Tr. is "ON"): L×IRP/toff=VF+Vout+Ronn×IoutEquation 4
Put Equation 4 to Equation 3 and solve for ON duty of P-channel transistor, Don=ton/(toff+ton), Don=(Vout-Ronn×Iout+RL×Iout)/(Vin+Ronn×Iout-Ronp×Iout)Equation 5
Ripple Current is as follows; IRP=(VIN-VOUT-RONP×IOUT-RL×IOUT)×DoN/fosc/LEquation 6
wherein, peak current that flows through L, and Lx Tr. is as follows; ILmax=lout+lrp/2Equation 7
Consider ILmax, condition of input and output and select external components.

★The above explanation is directed to the calculation in an ideal case in continuous mode.

How to Adjust Output Voltage and about Phase Compensation

As for Adjustable Output type, feedback pin (V_{FB}) voltage is controlled to maintain 0.8V. Output Voltage, V_{OUT} is as following equation;

Vout R1+R2= V_{FB} :R2 Vout= V_{FB} ×(R1+R2)/R2

Thus, with changing the value of R1 and R2, output voltage can be set in the specified range.

In the DC/DC converter, with the load current and external components such as L and C, phase might be behind 180 degree. In this case, the phase margin of the system will be less and stability will be worse. To prevent this, phase margin should be secured with proceeding the phase. A pole is formed with external components L and Cout.

A zero (signal back to zero) is formed with R1 and Cb.

First, choose the appropriate value of R1, R2 and Cb. Set R1+R2 value $100k\Omega$ or less.

For example, if L=4.7 μ H, Cout=10 μ F, the cut off frequency of the pole is approximately 23kHz. To make the cut off frequency of the zero by R1, R2, and Cb be higher than 23kHz, set R1=33k Ω and Cb=100pF.If Vout is set at 2.0V, R2=22k Ω is appropriate.

External Components

1.Inductor

Select an inductor that peak current does not exceed ILmax. If larger current than allowable current flows, magnetic saturation occurs and makes transform efficiency be worse.

Supposed that the load current is at the same, the smaller value of L is used, the larger the ripple current is.

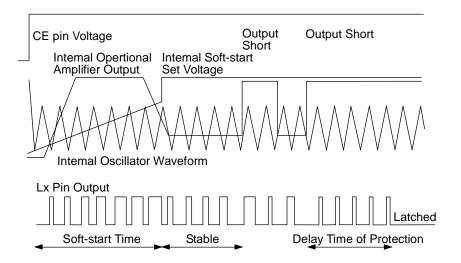
Provided that the allowable current is large in that case and DC current is small, therefore, for large output current, efficiency is better than using an inductor with a large value of L and vice versa.

2.Capacitor

As for C_{IN} , use a capacitor with low ESR (Equivalent Series Resistance) Ceramic type of a capacity at least $10\mu F$ for stable operation.

Cout can reduce ripple of the output voltage, therefore as much as 10µF ceramic type is recommended.

TIMING CHART



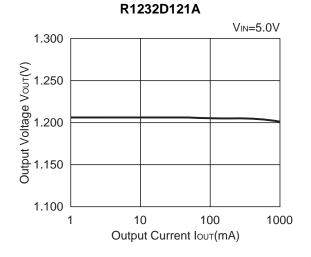
The timing chart as shown above describes the waveforms starting from the IC is enabled with CE and latched with protection. During the soft-start time, until the level is rising up to the internal soft-start set voltage, the duty cycle of Lx is gradually wider and wider to prevent the over-shoot of the voltage. During the term, the output of amplifier is "H". After the output voltage reaches the set output voltage, they are balanced well. Herein, if the output pin would be short circuit, the output of amplifier would become "H" again, and the condition would continue for 2.0ms (Typ.), or the Lx limit current is exceeded even once when the driver operates by duty 100%, latch circuit would work and the output of Lx would be latched with "OFF". (Output ="High-Z")

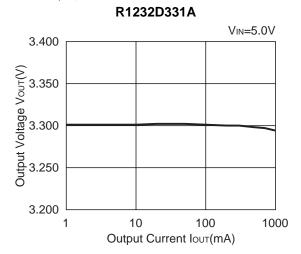
If the output short is released before the latch circuit works (within 2ms after output shorted), the output of amplifier is balanced in the stable state again.

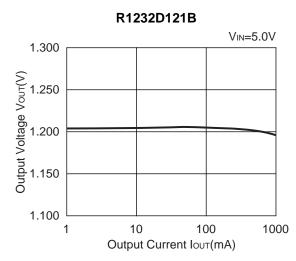
Once the IC is latched, to release the protection, input "L" with CE pin, or make the supply voltage at UVLO level or less.

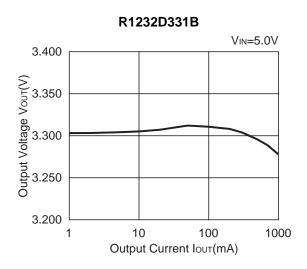
TYPICAL CHARACTERISTICS

1) Output Voltage vs. Output Current ($C_{IN} = 10 \mu F$, $C_{OUT} = 10 \mu F$)

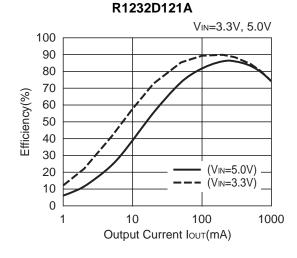


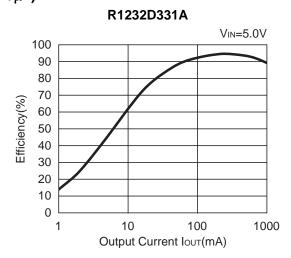


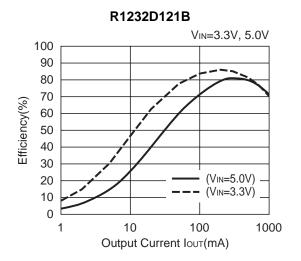


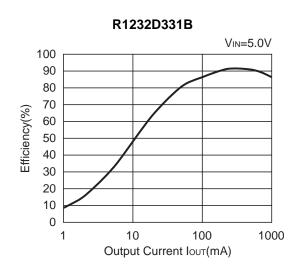


2) Efficiency vs. Output Current ($C_{IN} = 10 \mu F$, $C_{OUT} = 10 \mu F$)

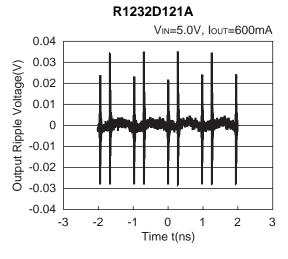


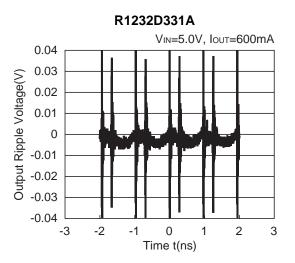


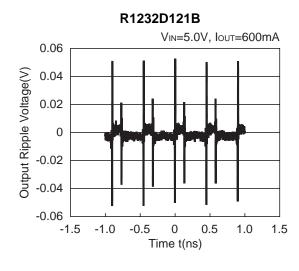


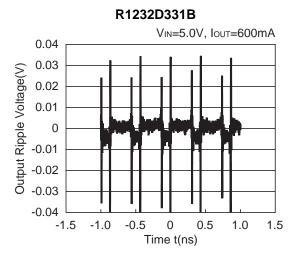


3) Output Waveform

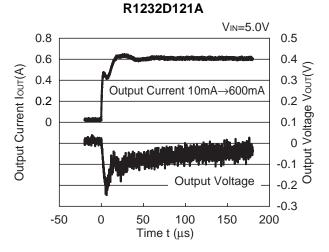


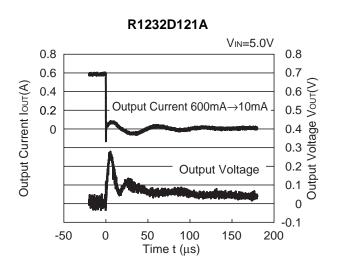




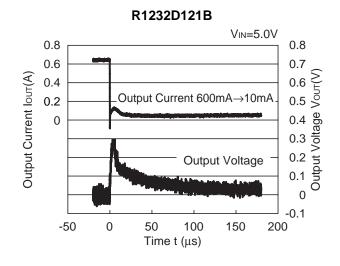


4) Load Transient Response

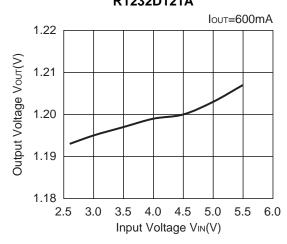


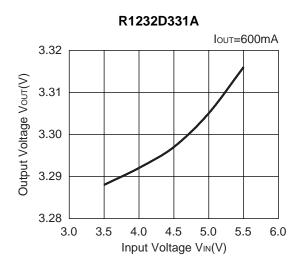


R1232D121B VIN=5.0V 0.5 8.0 0.4 0.6 Output Current lour(A) 0.4 Output Current 10mA→600mA 0.2 0 Output Voltage -0.2 -0.3 0 50 100 200 -50 150 Time t (µs)

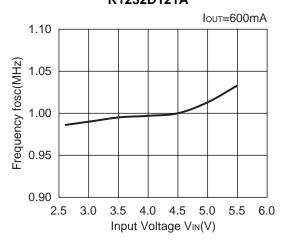


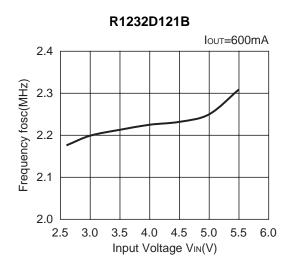
5) Output Voltage vs. Input Voltage R1232D121A



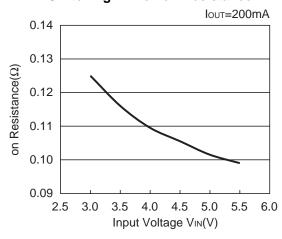


6) Oscillator Frequency vs. Input Voltage R1232D121A

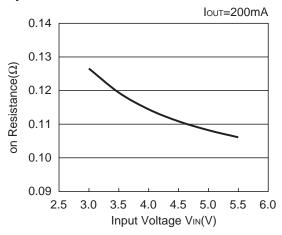




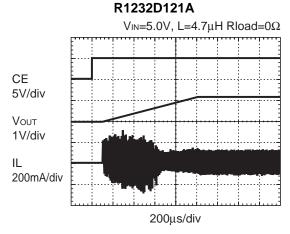
7) Lx Transistor On Resistance vs. Input Voltage Switching Tr. Pch on Resistance



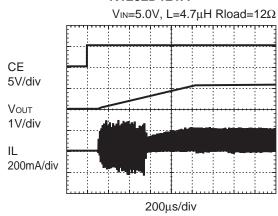
Synchronous Rectifier Tr. Nch on Resistance

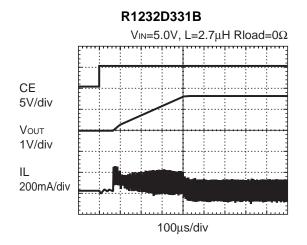


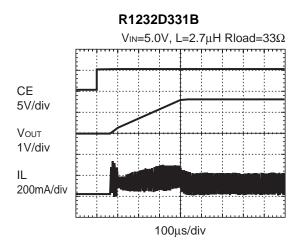
8) Turn-on speed by CE pin



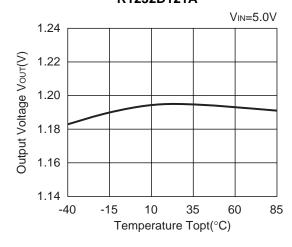
R1232D121A

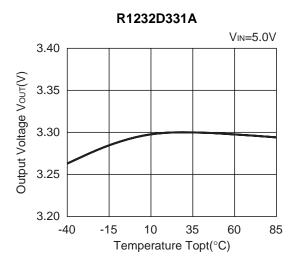




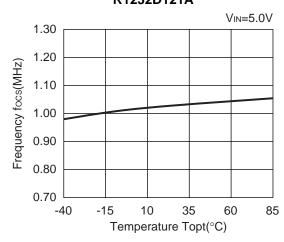


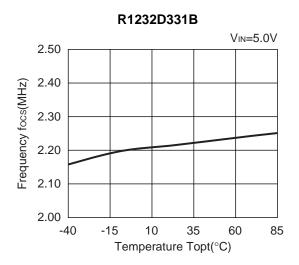
9) Output Voltage vs. Temperature R1232D121A



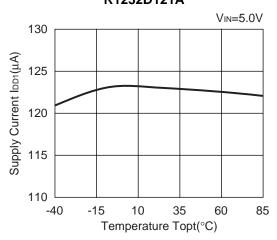


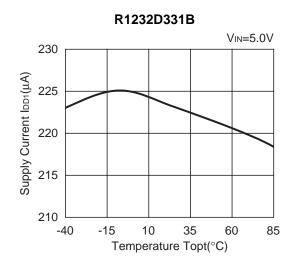
10) Oscillator Frequency vs. Temperature R1232D121A



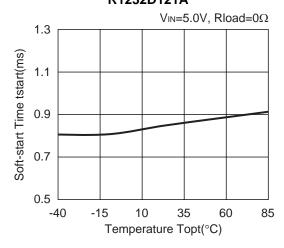


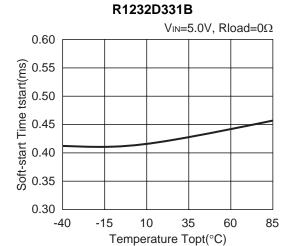
11) Supply Current vs. Temperature R1232D121A



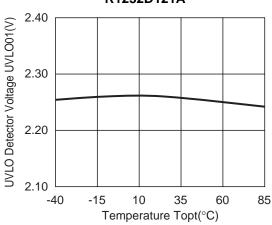


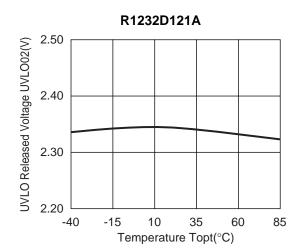
12) Soft-start time vs. Temperature R1232D121A



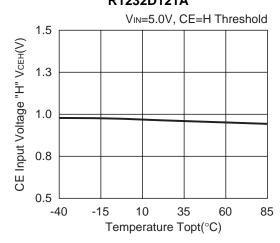


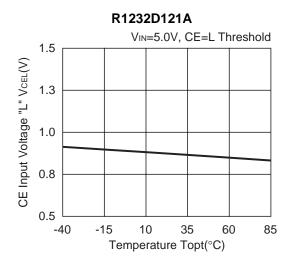
13) UVLO Voltage vs. Temperature R1232D121A



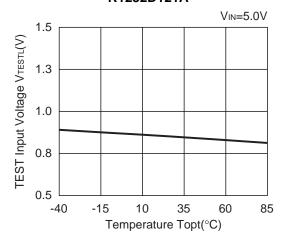


14) CE Input Voltage vs. Temperature R1232D121A

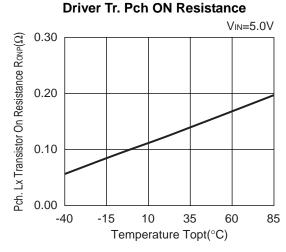




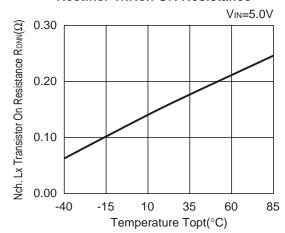
15) TEST Input Voltage vs. Temperature R1232D121A



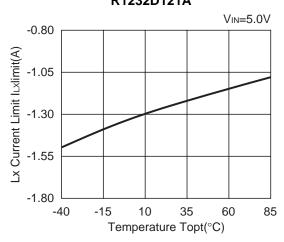
16) Lx Transistor On Resistance vs. Temperature

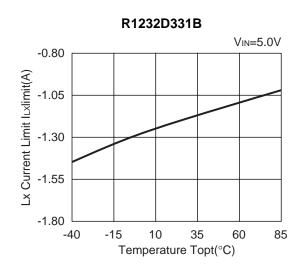


Rectifier Tr.Nch ON Resistance

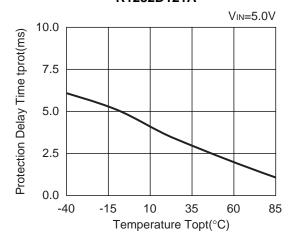


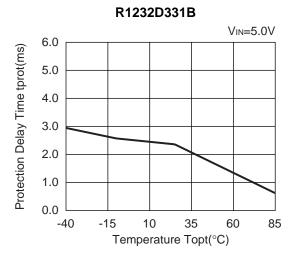
17) Current Limit vs. Temperature R1232D121A





18) Protection Delay Time vs. Temperatures R1232D121A







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