# RICOH

# 2.5 A PWM Step-up DC/DC Converter

NO.EA-278-180731

# OUTLINE

The R1213K is a low supply current PWM step-up DC/DC converter capable of providing an output current up to 2.5 A. Internally, the device consists of an Nch MOSFET driver, an oscillator, a PWM comparator, a voltage reference unit, an error amplifier, a soft-start circuit, an under voltage lockout circuit (UVLO), a thermal shutdown protection circuit, an overcurrent protection circuit and an latch-type protection circuit.

The R1213K requires minimal external component count. By simply using an inductor, resistors, capacitors and a diode, a high-efficiency step-up DC/DC converter can be easily configured.

The R1213K can adjust the output voltage, the soft-start time, the phase compensation using the external resistors and capacitors.

The R1213K has a shutdown control function which can be activated by a protection circuit to turn off the external Pch MOSFET for breaking the current path between the input and output.

The R1213K provides an overcurrent protection circuit, a latch-type protection circuit, a thermal shutdown protection circuit and an UVLO circuit. The overcurrent protection circuit limits the L<sub>x</sub> peak current and a latch-type protection circuit latches the Nch MOSFET off to stop the operation of the DC/DC converter if the output voltage drop due to overcurrent continues more than the protection delay time.

The R1213K is offered in a 12-pin DFN(PLP)2730-12 package.

# FEATURES

- Input Voltage Range (Maximum Rating) ..... 2.3 V to 5.5 V (6.5 V)
- Supply Current ...... Typ. 550 µA (non-switching)
- Supply Current ..... Typ. 3 mA (switching)
- Standby Current······ Max. 1.5 µA (CE = "L")
- Feedback Voltage Accuracy ······ ±8 mV
- Feedback Voltage Temperature Coefficient ..... ±50 ppm/°C
- - 500 mA: V<sub>IN</sub> = 2.3 V, V<sub>OUT</sub> = 5.0 V
    - 250 mA:  $V_{IN}$  = 2.7 V,  $V_{OUT}$  = 9.6 V
      - 150 mA: V<sub>IN</sub> = 3.0 V, V<sub>OUT</sub> = 15 V
- Nch ON Resistance ...... Typ. 0.07 Ω
- Shutdown Control Function ..... Activated by the external Pch MOSFET
- Thermal Shutdown Circuit ······ Activated at 150°C (Hys.= 40°C)
- Overcurrent Protection Circuit ······ Activated at Typ. 3.0 A

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- Latch-type Protection Circuit ..... Protection Delay Time: Typ. 32 ms
- FLAG Output Function ..... Activated at "H"
- UVLO Detector Threshold ..... Typ. 2.0 V
- Oscillator Frequency ..... Typ. 1.0 MHz
- Maximum Duty Cycle ..... Min. 85%, Typ. 90%
- Soft-start Time ...... Set by the SS Pin
- Phase Compensation
  Set by the AMPOUT Pin
- Package DFN(PLP)2730-12

# APPLICATION

- Flash LEDs
- Data Cards
- DSCs
- LCD Source Bias Supplies

# SELECTION GUIDE

The R1213K offers users to select the output voltage type matched to their set output voltage. Selecting the matched output voltage type can ensure high-speed transient response and stability.

#### **Selection Guide**

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
R1213K001*-TR	DFN(PLP)2730-12	5,000 pcs	Yes	Yes

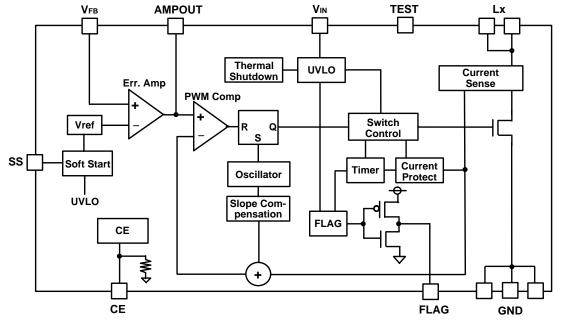
\*: Specify the output voltage type.

A: Low Output Voltage Type (Vout: 3.0 V to 6.0 V)

B: High Output Voltage Type (Vout: 6.0 V to 15 V)

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# **BLOCK DIAGRAMS**

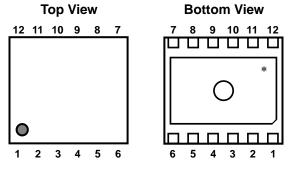


R1213K Block Diagram

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# **PIN DESCRIPTION**



DFN(PLP)2730-12 Pin Configuration

#### DFN(PLP)2730-12 Pin Description

Pin No	Symbol	Pin Description			
1	AMPOUT	Amplifier Output Pin			
2	Vfb	Feedback Voltage Pin			
3	CE	Chip Enable Pin, Active-high			
4	GND	Ground Pin <sup>(1)</sup>			
5	GND	Ground Pin <sup>(1)</sup>			
6	GND	Ground Pin <sup>(1)</sup>			
7	TEST	TEST Pin <sup>(2)</sup>			
8	Lx	Switching Pin <sup>(1)</sup>			
9	Lx	Switching Pin <sup>(1)</sup>			
10	Vin	Input Voltage Pin			
11	FLAG	Shutdown Control Pin <sup>(3)</sup>			
12	SS	Soft-start Pin			

\* The tab on the bottom of the package enhances thermal performance and is electrically connected to GND (substrate level). It is recommended that the tab be connected to the ground plane on the board, or otherwise be left floating.

<sup>&</sup>lt;sup>(1)</sup> The No.4, No.5 and No.6 pins must be connected together. The No.8 and No.9 pins must be connected together.

<sup>&</sup>lt;sup>(2)</sup> The TEST pin must be connected to GND or left floating.

<sup>&</sup>lt;sup>(3)</sup> The FLAG pin should be left floating when it is not used.

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# **ABSOLUTE MAXIMAM RATINGS**

bsolute Maxim	osolute Maximum Ratings			
Symbol	Item	Rating	Unit	
VIN	V <sub>IN</sub> Pin Voltage	-0.3 to 6.5	V	
Vampout	AMPOUT Pin Voltage	-0.3 to V <sub>IN</sub> + 0.3	V	
Vce	CE Pin Voltage	-0.3 to 6.5	V	
VFB	V <sub>FB</sub> Pin Voltage	-0.3 to 6.5	V	
Vss	SS Pin Voltage	-0.3 to V <sub>IN</sub> + 0.3	V	
V <sub>FLG</sub>	FLAG Pin Voltage	-0.3 to V <sub>IN</sub> + 0.3	V	
VTST	TEST Pin Voltage	-0.3 to V <sub>IN</sub> + 0.3	V	
V <sub>LX</sub>	Lx Pin Voltage	-0.3 to 18.0	V	
PD	Power Dissipation <sup>(1)</sup> (DFN(PLP)2730-12, JEDEC STD. 51-7 Test Land Pattern)	3100	mW	
Tj	Junction Temperature Range	-40 to 125	°C	
Tstg	Storage Temperature Range	-55 to 125	°C	

#### ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings are not assured.

# RECOMMENDED OPERATING CONDITIONS

Symbol	Item	Rating	Unit
V <sub>IN</sub>	Input Voltage	2.3 to 5.5	V
Та	Operating Temperature Range	-40 to 85	°C

#### **RECOMMENDED OPERATING CONDITIONS**

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such ratings by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

<sup>(1)</sup> Refer to POWER DISSIPATION for detailed information.

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# **ELECTRICAL CHARACTERISTICS**

Symbol	ltem	Conditions	Min.	Тур.	Max.	Unit
Istandby	Standby Current	$V_{IN} = 5.5 V, V_{CE} = 0 V$		0.1	1.5	μA
IDD1	Supply Current 1 (non-switching)	$V_{IN} = 5.5 \text{ V}, \text{ Vfb} = 0.9 \text{ V}$		550	800	μA
I <sub>DD2</sub>	Supply Current 2	$V_{IN} = 5.5 \text{ V}, \text{ V}_{FB} = 0 \text{ V}$		3.0	4.5	mA
VUVLO1	UVLO Detector Threshold	V <sub>FB</sub> = 0 V	1.9	2.0	2.1	V
V <sub>UVLO2</sub>	UVLO Released Voltage	V <sub>FB</sub> = 0 V		V <sub>UVLO1</sub> +0.12	2.25	V
Vout	Output Voltage Range	R1213K001A	3.0		6.0	V
VOUT		R1213K001B	6.0		15	v
VFB	Feedback Voltage Accuracy	V <sub>IN</sub> = 3.6 V	0.792	0.8	0.808	V
∆V <sub>FB</sub> /∆Ta	Feedback Voltage Temperature Coefficient	-40°C ≤ Ta ≤ 85°C		±50		ppm /°C
ILXLEAK	Lx Leakage Current	$V_{LX} = 16 V, V_{CE} = 0 V$		0.01	2.0	μA
<b>I</b> FBH	V <sub>FB</sub> "H" Input Current	$V_{IN} = 5.5 \text{ V}, \text{ V}_{FB} = 5.5 \text{ V}$			0.15	μA
FBL	VFB "L" Input Current	$V_{IN} = 5.5 \text{ V}, \text{ V}_{FB} = 0 \text{ V}$	-0.15			μA
ICEL	V <sub>CEL</sub> Input Current	$V_{IN} = 5.5 \text{ V}, \text{ V}_{CE} = 0 \text{ V}$	-0.2		0.2	μA
RCE	CE Pull-down Resistance			1000		kΩ
lss	Soft-start Current	V <sub>IN</sub> = 3.6 V		10		μA
VCEH	CE Input Voltage "H"	V <sub>IN</sub> = 5.5 V	1.5			V
VCEL	CE Input Voltage "L"	V <sub>IN</sub> = 2.3 V			0.3	V
fosc	Oscillator Frequency	$V_{IN} = 3.6 \text{ V}, \text{ V}_{FB} = 0 \text{ V}$	0.85	1.00	1.15	MHz
Maxduty	Maximum Duty Cycle	$V_{IN} = 3.6 \text{ V}, \text{ V}_{FB} = 0 \text{ V}$	85	90	95	%
TTSD	Thermal Shutdown Temperature	Junction Temperature		150		°C
T <sub>TSR</sub>	Thermal Shutdown Released Temperature	Junction Temperature		110		°C
gm	Trans-conductance <sup>(1)</sup>	VIN = 3.6 V		220		μS
ILXLIM	Lx Current Limit	V <sub>IN</sub> = 3.6 V	2.5	3.0	3.8	А
Ron	Nch ON Resistance <sup>(1)</sup>	V <sub>IN</sub> = 3.6 V		0.07		Ω
tprot	Latch-type Protection Delay Time	V <sub>IN</sub> = 3.6 V		32		ms
RUSH	Inrush Current <sup>(2)</sup>				1.5	А

 $<sup>^{\</sup>left( 1\right) }$  Guaranteed by design engineering, not mass production tested.

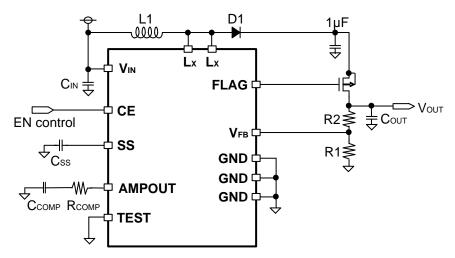
<sup>&</sup>lt;sup>(2)</sup> Guaranteed by design engineering when the external Pch MOSFET is connected to the FLAG pin. Refer to the recommended components at *APPLICATION INFORMATION* and *TECHNICAL NOTES*.

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# APPLICATION INFORMATION

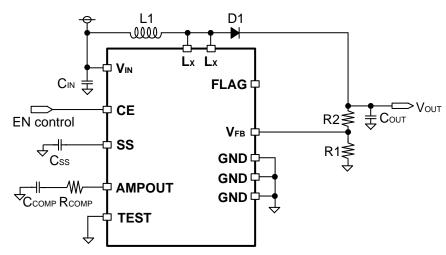
## **Typical Application**

External Pch MOSFET is Connected for Breaking the Current Path between VIN – VOUT (VOUT < 13 V)



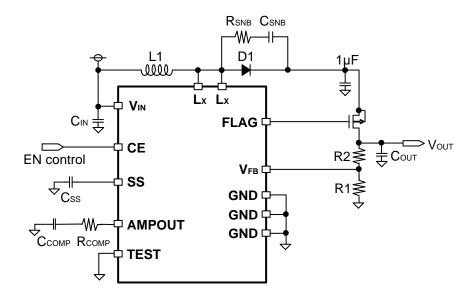
Notes: The GND pins and also the  $L_X$  pins must be mutually short-circuited right near the ground plane on the board. The TEST pin must be connected to the ground plane on the board or be left floating.

## External Pch MOSFET is <u>NOT</u> Connected for Breaking the Current Path between $V_{IN} - V_{OUT}$ ( $V_{OUT} < 13 V$ )



Notes: The GND pins and also the L<sub>x</sub> pins must be mutually short-circuited right near the ground plane on the board. The TEST pin must be connected to the ground plane on the board or be left floating. The FLAG pin must be left floating.

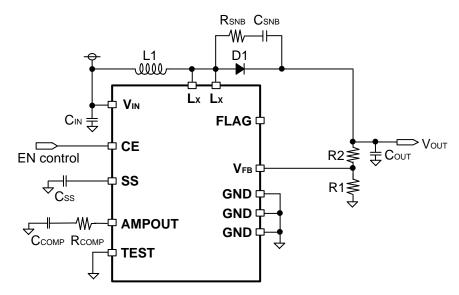
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External Pch MOSFET is Connected for Breaking the Current Path between  $V_{IN} - V_{OUT} (V_{OUT} \ge 13 \text{ V})$ 

Notes: The GND pins and also the L<sub>x</sub> pins must be mutually short-circuited right near the ground plane on the board. The TEST pin must be connected to the ground plane on the board or be left floating. The snubber circuit must be added for preventing spike noise on the L<sub>x</sub> pin.

#### External Pch MOSFET is <u>NOT</u> Connected for Breaking the Current Path between V<sub>IN</sub> − V<sub>OUT</sub> (V<sub>OUT</sub> ≥ 13 V)



Notes: The GND pins and also the L<sub>x</sub> pins must be mutually short-circuited right near the ground plane on the board. The TEST pin must be connected to the ground plane on the board or be left floating. The FLAG pin must be left floating. The snubber circuit must be added for preventing spike noise on the L<sub>x</sub> pin.

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	V <sub>IN</sub>	Cap.	Spec.	Part Name	Manufacturer
CIN	All	10 µF	6.3 V	C2012JB0J106M	TDK
	V <sub>OUT</sub>	Cap.	Spec.	Part Name	Manufacturer
	≤ 5 V	10 µF	6.3 V	C2012JB0J106M	TDK
0	≤ 10 V	10 µF	16 V	C2012X5R1C106K	TDK
Соит	all	10 µF	25 V	C3216X5R1E106K	TDK
	all	10 µF	25 V	TMK325BJ106MN	Taiyo Yuden
				·	·
	Vout	Sp	ec.	Part Name	Manufacture
D4	all	40 V	′, 3 A	CMS16	TOSHIBA
D1	all	40 V	′, 3 A	RB056L-40	ROHM
	-				
	Vout	Ind.	Spec.	Part Name	Manufacturer

## **Recommended Components**

	Vout	Ind.	Spec.	Part Name	Manufacturer	
			2.2 A	SPM3012T-2R2N	TDK	
	$3.0V \le V_{OUT} \le 4.5V$	2.2 µH	2.7 A	SPM4012T-2R2N	TDK	
		μι	3.5 A	NR5040T2R2N	Taiyo Yuden	
L <b>1</b> <sup>(1)</sup>		4.7	1.7 A	SPM4012T-4R7N	TDK	
	4.5V < V <sub>OUT</sub> ≤ 12V	μH	3.1 A	NR5040T4R7N	Taiyo Yuden	
	$12V < V_{OUT} \le 15V$			1.4 A	VLF5014ST-6R8N	TDK
		2V < V <sub>OUT</sub> ≤ 15V 6.8 μH	2.8 A	RLF7030T-6R8N	TDK	
		P	3.7 A	NR8040T6R8N	Taiyo Yuden	

	Vout	Spec. (I <sub>DS</sub> , V <sub>DS</sub> , V <sub>GS</sub> )	Part Name	Manufacturer
Pch.MOSFET	all	4.5 A, −30 V, ±20 V	UPA1914	Renesas

<sup>&</sup>lt;sup>(1)</sup> It is recommended that the rated current of the inductor be higher than the LX limit current. Performing the current limitation outside of the R1213K requires the use of small components.

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## • Selection of Resistors and Capacitors for Phase Compensation

The R1213x requires an external phase compensation on the feedback loop for output voltage control to prevent the large output ripple, the unstable operation and the deterioration of device efficiency. Connect a resistor (RCOMP) and a capacitor (CCOMP) between the AMPOUT and GND pins. RCOMP and CCOMP can be calculated as follows:

## [R1213K001A]

$$\begin{split} R_{\text{COMP}} &= 90 \text{ x Vin x Vout x Cout / (L x Ioutmax)} \\ C_{\text{COMP}} &= 30 \text{ x Vout x L x Ioutmax / (VIN <math>^2 \text{ x R_{COMP}})} \end{split}$$

## [R1213K001B]

 $\begin{aligned} & \mathsf{R}_{\mathsf{COMP}} = 45 \text{ x Vin x Vout x Cout / (L x Ioutmax)} \\ & \mathsf{C}_{\mathsf{COMP}} = 30 \text{ x Vout x L x Ioutmax / (Vin <sup>2</sup> x Rcomp)} \end{aligned}$ 

The appropriate values for  $R_{COMP}$  and  $C_{COMP}$  vary depending on the peripheral components and circuit board. Determine the appropriate values for  $R_{COMP}$  and  $C_{COMP}$  according to the transient response.

VIN (V)	Vout (V)	Ioutmax (mA)	C <sub>IN</sub> (µF)	Соит (µF)	L1 (µH)	D1	R <sub>COMP</sub> (kΩ)	C <sub>COMP</sub> (nF)
3.3	3.8	1200	10	20	2.2	3 A	8.2	3.3
3.3	5	800	10	20	4.7	3 A	8.2	6.8
3.3	12	250	10	20	4.7	3 A	27	1.8
5.0	15	650	10	20	6.8	3 A	15	5.1

# Output Voltage Setting

The output voltage can be calculated by the values of resistors (R1 and R2) as follows:

Output Voltage = 
$$V_{FB} x (R1 + R2) / R1$$
  
( $V_{FB} = 0.8 V$ )

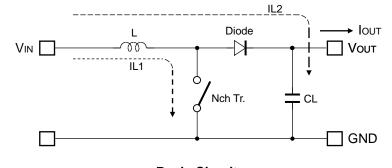
Notes: Set the sum of R1 and R2 to be 200  $k\Omega$  or less.

# Soft-start Time Setting

The soft-start time can be adjusted by a capacitor ( $C_{SS}$ ) between the SS and GND pins. The soft-start time can be calculated as follows:

Soft-start time =  $C_{SS} \times V_{FB} / I_{SS}$ = 8 x  $C_{SS} \times 10^4$  [sec] ( $V_{FB}$  = 0.8 V,  $I_{SS}$  = 10 µA)

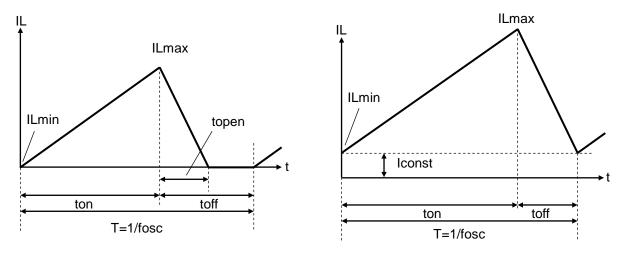
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## • Operation of Step-Up Dc/Dc Converter and Output Current



#### Current (IL) Flowing Through Inductor (L)



#### Discontinuous Inductor Current Mode

**Continuous Inductor Current Mode** 

The PWM control type of the step-up DC/DC converter has two operation modes characterized by the continuity of inductor current: discontinuous inductor current mode and continuous inductor current mode.

When an Nch transistor is in On-state, the voltage to be applied to the inductor (L) is described as  $V_{IN}$ . An increase in the inductor current (IL1) can be written as follows:

IL1 = V<sub>IN</sub> x ton / L ...... Formula 1

In the step-up DC/DC converter circuit, the energy accumulated during the On-state is transferred into the capacitor even in the Off-state. A decrease in the inductor current (IL2) can be written as follows:

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In the PWM control, IL1 and IL2 become continuous when topen = toff, which is called continuous inductor current mode.
When the device is in continuous inductor current mode and operates in steady-state conditions, the variations of IL1 and IL2 are same:
$V_{IN} x \text{ ton } / L = (V_{OUT} - V_{IN}) x \text{ toff } / L$
Therefore, the duty cycle in continuous inductor current mode is:
duty (%)= ton / (ton + toff) = (V <sub>OUT</sub> - V <sub>IN</sub> ) / V <sub>OUT</sub>
When topen = toff, the average of IL1 is:
IL1 (Ave.) = V <sub>IN</sub> x ton / (2 x L)
If the input voltage (V <sub>IN</sub> ) is equal to the output voltage (V <sub>OUT</sub> ), the output current ( $I_{OUT}$ ) is:
$I_{OUT} = V_{IN}^2 x \text{ ton } / (2 x L x V_{OUT})$ Formula 6
If IOUT is larger than Formula 6, the device switches to continuous inductor current mode
The L <sub>x</sub> peak current flowing through L (ILmax) is:
$ILmax = I_{OUT} \times V_{OUT} / V_{IN} + V_{IN} \times ton / (2 \times L) \dots$ Formula 7
ILmax = Iout x Vout / VIN + VIN x T x (Vout - VIN) / (2 x L x Vout) Formula 8

As a result, ILmax becomes larger compared to  $I_{OUT}$ . The overcurrent protection circuit operates if the ILmax becomes more than the  $L_X$  current limit. When considering the input and output conditions or selecting the external components, please pay attention to ILmax.

**Notes:** The above calculations are based on the ideal operation of the device. They do not include the losses caused by the external components or Nch transistor. The actual maximum output current will be 50% to 80% of the above calculation results. Especially, if IL is large or  $V_{IN}$  is low, it may cause the switching losses.

# **TECHNICAL NOTES**

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed a rated voltage, a rated current or a rated power. When designing a peripheral circuit, please be fully aware of the following points.

- Ensure that the V<sub>IN</sub> and GND lines are firmly connected. A large switching current flows through the V<sub>IN</sub> and GND lines. If their impedance is too high, noise pickup or unstable operation may result.
- When an Nch MOSFET driver is turned off, the inductor may generate a spike-shaped high voltage.
  Use a high-break-down-voltage capacitor (C<sub>OUT</sub>) and a high-break-down-voltage diode that are 1.5 times or more than the set output voltage.
- Choose a schottky diode (D1) that has low forward voltage, low reverse current, and is fast in switching speed.
- Use an inductor that has a low DC resistance, has an enough tolerable current and is less likely to cause magnetic saturation.
- The FLAG pin (Shutdown Control Pin) turns off the external Pch MOSFET to break the current path between V<sub>IN</sub> and V<sub>OUT</sub> during standby, UVLO, thermal shutdown and latch-type protection. Place a capacitor of 1 µF between the source of the external Pch MOSFET and GND to protect the external Pch MOSFET from overvoltage caused by the inductor current. During the soft-start, the FLAG pin turns on or off the external Pch MOSFET synchronizing with the switching of the Nch MOSFET to prevent the inrush current. Select the external Pch MOSFET with fast

switching speed (Approx. 100 ns) and small gate capacity (3 nF or less).

• The spike noise of L<sub>x</sub> should not exceed the absolute maximum rating. The spike noise of L<sub>x</sub> may exceed the absolute maximum ratings under  $V_{OUT} \ge 13$  V. To reduce the spike noise of L<sub>x</sub>, place a snubber circuit (R<sub>SNB</sub> and C<sub>SNB</sub> are connected in series) parallel to the diode (D1). A snubber circuit may also be required under  $V_{OUT} < 13$  V if the spike noise of L<sub>x</sub> is large. It is recommended that a capacitor (C<sub>SNB</sub>) be 1100 pF and a resistor (R<sub>SNB</sub>) be 0.68  $\Omega$ . The appropriate values for C<sub>SNB</sub> and R<sub>SNB</sub> vary significantly depending on the circuit board and affect the device efficiency. Actual circuit board testing is required.

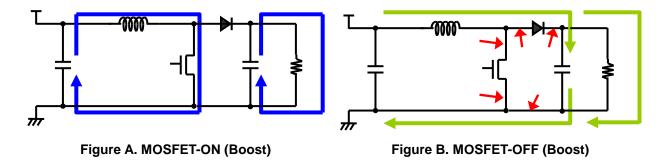
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 Latch-type protection circuit latches the Nch MOSFET off to stop the operation of the DC/DC converter if the output voltage drop due to overcurrent continues more than the protection delay time. When the latch-type protection circuit operates, the FLAG pin outputs "H" and turns the external Pch MOSFET off to break the current path between V<sub>IN</sub> and V<sub>OUT</sub>.

The protection delay time is set to typically 32 ms. If the output voltage returns to normal during the protection delay time, the internal timer will be reset.

To release the latch-type protection, set the CE pin "H" or make the power supply voltage lower than the UVLO detector threshold.

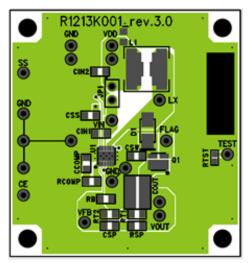
- Connect the TEST pin to GND or otherwise leave it floating.
- Connect the FLAG pin to the external Pch MOSFET gate only.
- To prevent inrush current, connect the SS pin to a capacitor (Css) only.
- The tab on the bottom of the package enhances thermal performance and is electrically connected to GND (substrate level). It is recommended that the tab be connected to the ground plane on the board, or otherwise be left floating. To enhance the thermal performance of multilayer circuit board, provide a thermal via under the tab on the bottom of the package.
- In Fig. A and Fig. B, the current paths on the boost DC/DC converter are shown. The current paths when the MOSFET turns on are shown in Fig. A, and the current paths when the MOSFET turns off are shown in Fig. B. The pointed parts with red arrows in Fig. B are where the current flows only when the MOSFET turns on, or off. The parasitic impedance, inductance, or parasitic capacitance of these parts have some impact on the stability of DC/DC converter, and may cause a noise generation. Therefore the parasitic impedance, capacitance, inductance must be as small as possible. Furthermore, the current paths shown in Fig. A and Fig. B must be as short as possible and as wide as possible.



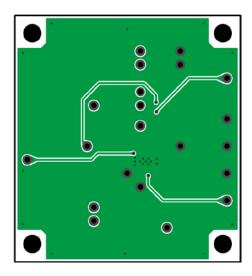
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## • PCB Layout

R1213K001A/B (PKG: DFN(PLP)2730-12pin)



Typical Board Layout – Top Layer



Typical Board Layout – Back Layer Note: R2 patterns are the layout for 2 serial resistance chips, RT1 and RT2 to set preferred value easier.

# **RICOH**

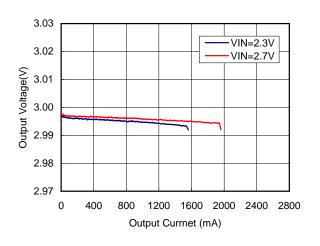
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# **TYPICAL CHARACTERISTICS**

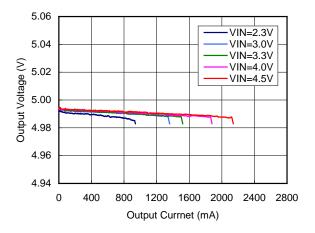
Note: Typical Characteristics are intended to be used as reference data; they are not guaranteed.

## 1) Output Voltage vs. Output Current (Ta = 25°C)

 $V_{OUT} = 3.0 V$ 

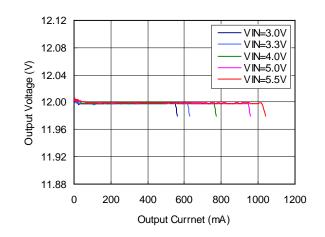


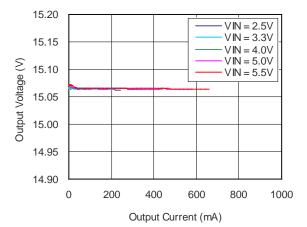
 $V_{OUT} = 5.0 V$ 





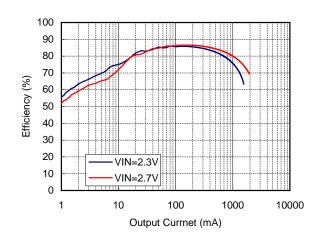






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2) Efficiency vs. Output Current (Ta = 25°C)  $V_{OUT} = 3.0 V$ 



Vout = 12 V

0.5

0.4

0.3

0.2

0.1

-0.1

0

-50

-25

0

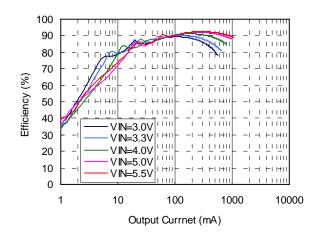
25

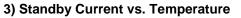
Temperature (°C)

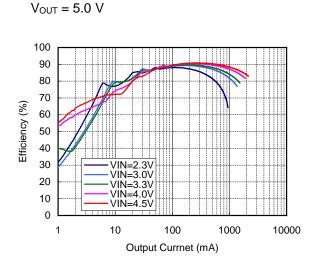
50

75

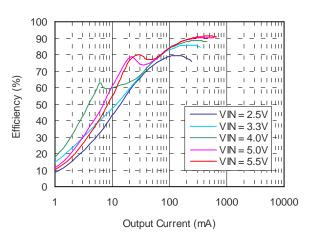
Standby Current (uA)



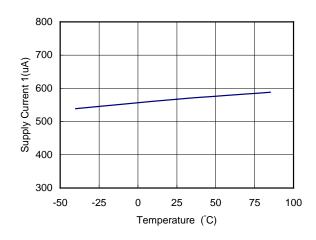




Vout = 15 V





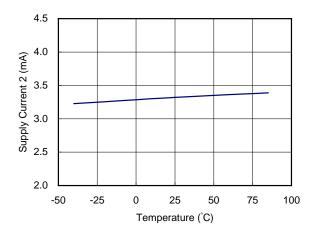


RICOH

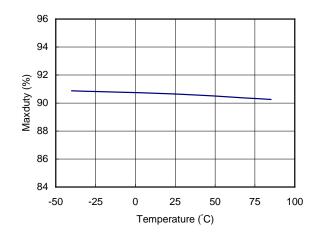
100

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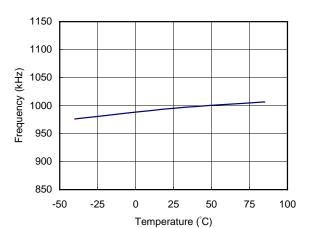
#### 5) Supply Current 2 vs. Temperature



#### 7) Maxduty vs. Temperature

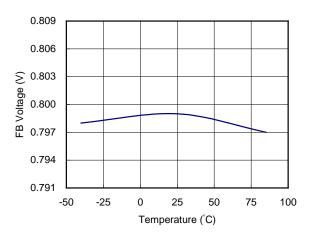




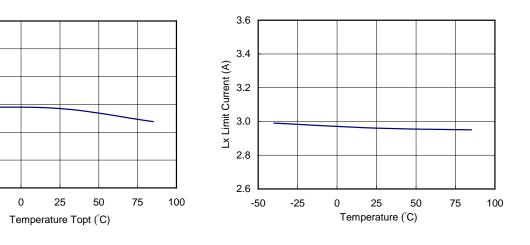


6) Frequency vs. Temperature





10) Lx Limit Current vs. Temperature



**RICOH** 

1.5

1.3

1.1

0.9

0.7

0.5

0.3

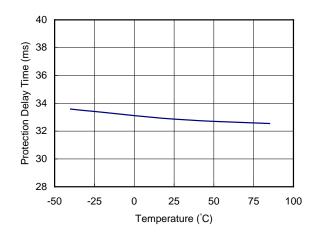
-50

-25

CE "H" Voltage (V)

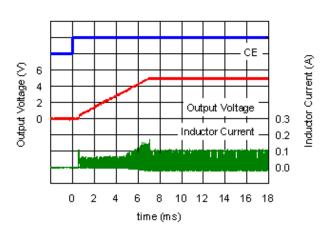
#### <u>R1213K</u>

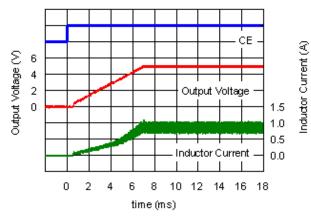
NO.EA-278-180731

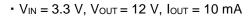


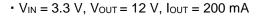
#### 11) Protection Delay Time vs. Temperature

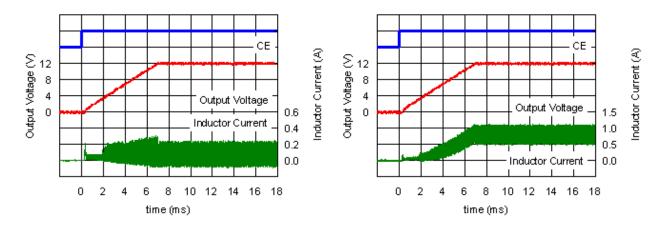
12) Start-up Waveform (Ta = 25°C, C<sub>SS</sub> = 0.1 μF, External Pch MOSFET Connected between V<sub>IN</sub> – V<sub>OUT</sub>) • V<sub>IN</sub> = 3.3 V, V<sub>OUT</sub> = 5.0 V, I<sub>OUT</sub> = 10 mA • V<sub>IN</sub> = 3.3 V, V<sub>OUT</sub> = 5.0 V, I<sub>OUT</sub> = 500 mA











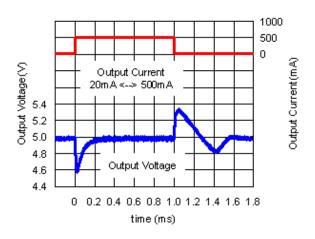
**RICOH** 

NO.EA-278-180731

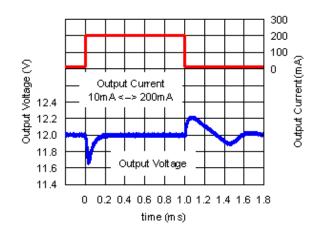
#### 13) Load Transient Response Waveform (Ta = 25°C)

•  $V_{IN} = 3.3 \text{ V}, V_{OUT} = 5.0 \text{ V}, I_{OUT} = 20 \Leftrightarrow 500 \text{ mA}$ L = 4.7 µH, C<sub>OUT</sub> = 20 µF, R<sub>COMP</sub> = 8.2 kΩ,

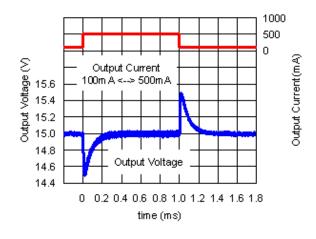
 $C_{\text{COMP}} = 6.8 \text{ nF}$ 



•  $V_{IN} = 3.0 \text{ V}, V_{OUT} = 12 \text{ V}, I_{OUT} = 10 \Leftrightarrow 200 \text{ mA}$   $L = 4.7 \text{ }\mu\text{H}, C_{OUT} = 20 \text{ }\mu\text{F}, R_{COMP} = 27 \text{ }k\Omega,$  $C_{COMP} = 1.8 \text{ nF}$ 

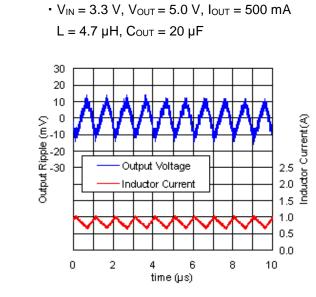


• V<sub>IN</sub> = 5.0 V, V<sub>OUT</sub> = 15.0 V, I<sub>OUT</sub> = 100 ⇔ 500 mA L = 6.8 µH, C<sub>OUT</sub> = 20 µF, R<sub>COMP</sub> = 15 kΩ, C<sub>COMP</sub> = 5.1 nF



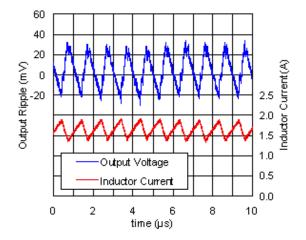
#### <u>R1213K</u>

NO.EA-278-180731

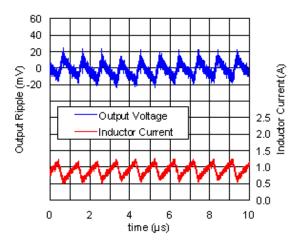


14) Output Voltage Waveform (Ta = 25°C)

- V<sub>IN</sub> = 5.0 V, V<sub>OUT</sub> = 15 V, I<sub>OUT</sub> = 500 mA L = 6.8  $\mu$ H, C<sub>OUT</sub> = 20  $\mu$ F



•  $V_{IN} = 3.3 \text{ V}, V_{OUT} = 12 \text{ V}, I_{OUT} = 200 \text{ mA}$ L = 4.7 µH, C<sub>OUT</sub> = 20 µF



# **POWER DISSIPATION**

# DFN(PLP)2730-12

Ver. A

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

#### **Measurement Conditions**

Item	Measurement Conditions
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	Outer Layer (First Layer): Less than 95% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100% of 50 mm Square
Through-holes	φ 0.3 mm × 23 pcs

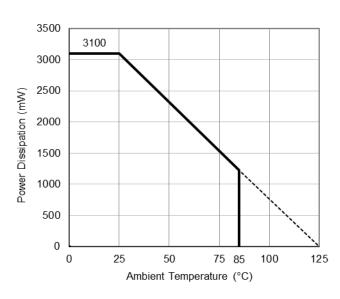
#### **Measurement Result**

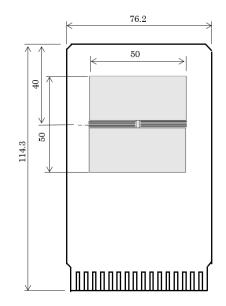
(Ta = 25°C, Tjmax = 125°C) ltem **Measurement Result Power Dissipation** 3100 mW Thermal Resistance (0ja)  $\theta ja = 32^{\circ}C/W$ Thermal Characterization Parameter (ψjt)  $\psi jt = 8^{\circ}C/W$ 

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θja: Junction-to-Ambient Thermal Resistance

wit: Junction-to-Top Thermal Characterization Parameter





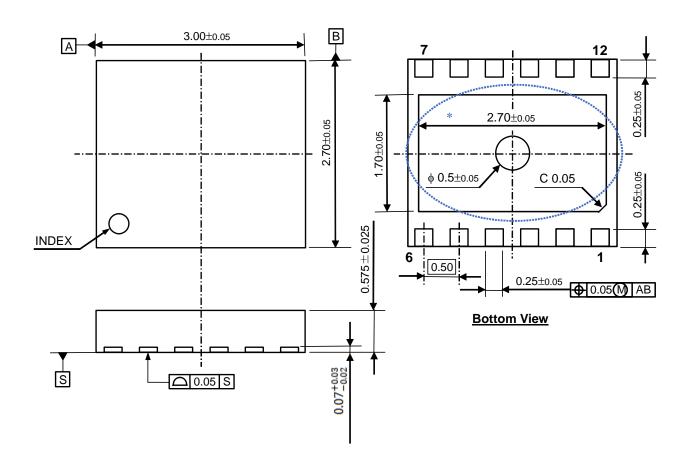
#### **Power Dissipation vs. Ambient Temperature**

**Measurement Board Pattern** 

# PACKAGE DIMENSIONS

# DFN(PLP)2730-12

Ver. A



DFN(PLP)2730-12 Package Dimensions (Unit: mm)

<sup>\*</sup>The tab on the bottom of the package shown by blue circle is a substrate potential (GND). It is recommended that this tab be connected to the ground plane on the board but it is possible to leave the tab floating.



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